Efficient Encoding Algorithm for Third-Order Spectral-Null Codes^{*}

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Abstract

An efficient algorithm is presented for encoding unconstrained information sequences into a third-order spectral-null code of length n and redundancy $9 \log_2 n + O(\log \log n)$. The encoding can be implemented using O(n) integer additions and $O(n \log n)$ counter increments.

Keywords: DC-free codes, spectral-null codes.

1 Introduction

Let F be the bipolar alphabet $\{+1, -1\}$. A word $\underline{x} = (x_1, x_2, \ldots, x_n)$ in F^n is a k-th order spectral-null word (at zero frequency) if the respective real polynomial $x_1z + x_2z^2 + \ldots + x_nz^n$ is divisible by $(z-1)^k$. We denote by $\mathcal{S}(n,k)$ the set of all k-th order spectral-null words in F^n . Any subset \mathcal{C} of $\mathcal{S}(n,k)$ is called a k-th order spectral-null code of length n. The concatenation of any l words in \mathcal{C} yields a word in $\mathcal{S}(nl,k)$; so, spectral-null codes can be used as block codes with a redundancy of $n - \log_2 |\mathcal{C}|$ bits (per block of length n).

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The set $\mathcal{S}(n,k)$ is equivalently characterized by

$$\mathcal{S}(n,k) = \left\{ \underline{x} \in F^n : \sum_{j=1}^n (j+c)^\ell x_j = 0, \quad \ell = 0, 1, \dots, k-1 \right\},$$
(1)

where c is any real constant (see [5], [6, Ch. 9]).

First-order spectral-null codes are also known by the names balanced codes, zero-disparity codes, or *DC-free codes*. There are known efficient encoding algorithms for these codes due to Knuth [3], Al-Bassam and Bose [1], and Tallini, Capocelli, and Bose [8]. Those algorithms result in codes with redundancy at most $\lceil \log_2 n \rceil$, where *n* is the code length. By 'efficient' we refer to the time and space complexity of the encoding; for example, in one of Knuth's algorithms, the complexity amounts to a look-up table of $\lceil \log_2 n \rceil^2$ bits and O(n)increments/decrements of a $\lceil \log_2 n \rceil$ -bit counter (as shown in [3], the space requirement can be eliminated by increasing the redundancy to $\log_2 n + O(\log \log n)$). The redundancy of S(n, 1) is $\frac{1}{2} \log_2 n + O(1)$, and such redundancy can be attained by enumerative coding [6, p. 117]. In terms of complexity, however, enumerative coding is less efficient than Knuth's algorithms or the algorithms in [1] and [8].

Efficient coding algorithms for the second-order spectral-null case were presented in [5] and [7]. Those algorithms have redundancy of $3 \log_2 n + O(\log \log n)$ bits and time complexity which amounts to O(n) additions of $O(\log n)$ -bit integers. Enumerative coding already turns out to be impractical for this case [5]. The redundancy of $\mathcal{S}(n, 2)$ is known to be $2 \log_2 n + O(1)$ [7].

For higher orders k of spectral null, Karabed and Siegel presented in [2] a coding method based upon finite-state diagrams (see also Monti and Pierobon [4]). However, since the rate of their construction is strictly less than 1, the resulting redundancy is linear in the code length n. It follows that for any fixed k and sufficiently large n, this redundancy is significantly larger than the upper bound $O(2^k \cdot \log n)$ on the redundancy of S(n,k); this bound is proved in [5] by nonconstructive arguments. A recursive construction is presented in [5] whose redundancy is $O(n^{1-\epsilon_k})$, where $0 < \epsilon_k < 1$ and $\lim_{k\to\infty} \epsilon_k = 0$. Yet, this redundancy is still considerably larger than the actual redundancy of S(n,k).

In this work, we present an efficient algorithm for encoding unconstrained sequences into a third-order spectral-null code whose redundancy is logarithmic in the code length. More specifically, for code length n, the redundancy is $9 \log_2 n + O(\log \log n)$ bits and the encoding complexity is O(n) additions of $O(\log n)$ -bit integers and $O(n \log n)$ increments/decrements of $\lceil \log_2 n \rceil$ -bit counters.

2 A third-order spectral-null encoder

It was shown in [5] that the length n of a third-order spectral-null word is divisible by 4, so we can write n = 2h for some even integer h. We will use the definition of $\mathcal{S}(2h,3)$ that is obtained from (1) by substituting k = 3 and c = -h-1. It will also be convenient hereafter to index the entries of a real word \underline{x} of length 2h by $(x_{-h}, x_{-h+1}, \ldots, x_{h-1})$. We define the moments of such a word \underline{x} by

$$\sigma_{\ell}(\underline{x}) \stackrel{\text{def}}{=} \sum_{j=-h}^{h-1} j^{\ell} \cdot x_j , \quad \ell = 0, 1, 2, \cdots .$$

Clearly, a word $\underline{x} \in F^n$ is in $\mathcal{S}(2h,3)$ if and only if $\sigma_0(\underline{x}) = \sigma_1(\underline{x}) = \sigma_2(\underline{x}) = 0$.

The following is an outline of our encoding algorithm. Let n = 2h where h is even and let m be the integer $\lceil \log_2 n \rceil = 1 + \lceil \log_2 h \rceil$. The input to the algorithm is a balanced word \underline{y} over F of length 2h - 6m + 2; namely, \underline{y} is a word in $\mathcal{S}(2h - 6m + 2, 1)$ that is generated from the raw data by any known DC-free encoder (e.g., [1], [3], or [8]). Our algorithm regards \underline{y} as a subword of a word \underline{x} of length n over $F \cup \{0\}$, where the remaining entries of \underline{x} are initially set to zero; hence, $\sigma_0(\underline{x}) = 0$. Next, the algorithm reduces to zero the absolute values of $\sigma_2(\underline{x})$ and $\sigma_1(\underline{x})$ (in that order), by a sequence of bit shifts and bit swaps, and by assigning values of F to the zero entries of \underline{x} . At this point, \underline{x} becomes a word in $\mathcal{S}(2h, 3)$. The encoding ends by coding recursively certain counters that were computed in the course of the algorithm, resulting in a word $\underline{x}' \in \mathcal{S}(2m + O(\log m), 3)$. The concatenation of \underline{x} and \underline{x}' , in turn, will form the output third-order spectral-null word.

The algorithm makes use of the following index sets, all being subsets of $S = \{-h, -h+1, \ldots, h-1\}$:

•
$$S_{\text{B2}} = \{d_i\}_{i=0}^{2m-8} \cup \{e_i\}_{i=0}^{2m-8}$$
, where

$$(d_i, e_i) = \begin{cases} (-10 \cdot 2^{i/2}, -6 \cdot 2^{i/2}) & \text{if } i \text{ is even} \\ (-9 \cdot 2^{(i+1)/2}, -7 \cdot 2^{(i+1)/2}) & \text{if } i \text{ is odd} \end{cases}, \quad 0 \le i \le 2m - 10, \quad (2)$$

 $(d_{2m-9}, e_{2m-9}) = (\tau_1, \tau_2)$, and $(d_{2m-8}, e_{2m-8}) = (-\tau_1, 7)$, where τ_1 is the smallest odd integer in S that is at least $\sqrt{(h^2/2) + 49}$, and τ_2 is the largest odd integer in S that is at most h/2. We remove $\{d_i, e_i\}$ from S_{B2} if $d_i < -h$.¹

• $S_{B3} = \{0, -3, 3, -5, 5, 6, -7, -9, 9, 10, -11, 12, -13, 14\}.$

• $S_{\rm C} = \{\pm 2^i\}_{i=0}^{m-2}$.

¹This can happen only for i = 2m-10, 2m-11. Nevertheless, in those cases where only $\{d_{2m-10}, e_{2m-10}\}$ can be removed, then $\{d_{2m-9}, e_{2m-9}\}$ is redundant as well. In fact, it turns out that we will need all the 2(2m-7) elements of S_{B2} only when h is close in value to a power of 2.

We will assume hereafter that h is large enough, in which case the sets S_{B2} , S_{B3} , and S_{C} are pairwise disjoint.² We let S_0 be the union $S_{B2} \cup S_{B3} \cup S_{C}$. Note that $|S_0| \leq 2(2m-7) + 14 + 2(m-1) = 6m-2$.

For a word \underline{x} of length n and a subset Y of S, we will use the notation $\langle \underline{x} \rangle_Y$ for the subword of \underline{x} that is indexed by Y.

The algorithm is summarized in Figure 1. The input y is of length $|S \setminus S_0| \ge 2h - 6m + 2$.

Step A: Initialization of \underline{x}

Let $\langle \underline{x} \rangle_{S \setminus S_0} \leftarrow$ balanced y. Let $\langle \underline{x} \rangle_{S_0} \leftarrow \underline{0}$.

Step B: Reduction of $|\sigma_2(\underline{x})|$

- **Step B1:** Shift cyclically the entries of $\langle \underline{x} \rangle_{S \setminus S_0}$, until the resulting \underline{x} is such that $|\sigma_2(\underline{x})| \leq h^2$. Let j_B be the smallest number of shifts applied until this condition is met.
- **Step B2:** For decreasing values of $i = 2m 8, 2m 9, \ldots, 0$, reduce the value of $|\sigma_2(\underline{x})|$ by assigning $x_{d_i} = -x_{e_i} = -1$ if $\sigma_2(\underline{x}) \ge 0$ and $x_{d_i} = -x_{e_i} = 1$ otherwise.
- **Step B3:** Let $\langle \underline{x} \rangle_{S_{B3}} \leftarrow$ the row in Table 1 that corresponds to $|\sigma_2(\underline{x})|$. If $\sigma_2(\underline{x}) \geq 0$ then let $\langle \underline{x} \rangle_{S_{B3}} \leftarrow -\langle \underline{x} \rangle_{S_{B3}}$ (i.e., negate $\langle \underline{x} \rangle_{S_{B3}}$).

Step C: Reduction of $|\sigma_1(\underline{x})|$

- **Step C1:** For increasing values of indexes $j = 1, 2, ..., \text{ swap } x_j$ with x_{-j} until $|\sigma_1(\underline{x})| \leq 2(h-1)$, and let j_C denote the number of swaps made until this condition is met.
- **Step C2:** For decreasing values of $i = m-2, m-3, \ldots, 0$, reduce the value of $|\sigma_1(\underline{x})|$ by assigning $x_{2^i} = -x_{-2^i} = -1$ if $\sigma_1(\underline{x}) \ge 0$ and $x_{2^i} = -x_{-2^i} = 1$ otherwise.

Step D: Recursive encoding

Apply Step A–C recursively to the binary representation of $(j_{\rm B}, j_{\rm C})$. Concatenate the resulting word, \underline{x}' , with \underline{x} to generate the final output of the encoder.

Figure 1: Third-order spectral-null encoder.

3 Analysis of the algorithm

3.1 Validity

We verify step by step that the algorithm indeed terminates with a third-order spectral-null word.

²As we show in the example of Section 4 and as pointed out in the previous footnote, some elements in S_{B2} may sometimes be excluded. This allows to have h as small as 18.

$\sigma_2(\underline{x}) \setminus \text{index}$	0	-3	3	-5	5	6	-7	-9	9	10	-11	12	-13	14
1	+	+	-	- 1	+	-	+	-	-	+	-	+	-	+
3	+	+	-	+	+	-	-	-	-	+	-	-	+	+
5	—	+	-	-	—	+	+	+	+	+	-	—	+	—
7	—	—	-	+	+	-	+	+	+	-	+	+	-	—
9	+	-	-	-	+	+	+	-	-	+	+	-	-	+
11	+	+	-	+	-	-	-	+	-	+	-	+	+	-
13	+	+	-	-	+	-	+	-	+	-	-	-	+	+
15	-	-	-	+	+	-	+	+	+	+	-	-	+	-
17	+	-	-	-	-	+	+	+	-	+	+	+	-	-
19	—	—	-	+	+	+	-	+	—	+	+	+	-	—
21	+	+	-	-	-	-	+	+	+	-	-	+	+	-
23	+	+	-	-	—	+	+	-	—	+	-	+	-	+
25	+	+	-	-	+	+	-	-	—	+	-	—	+	+
27	+	+	—	-	—	-	-	+	+	+	+	+	-	—
29	—	—	—	+	—	+	+	+	+	-	+	+	—	—
31	—	—	—	+	+	+	-	+	+	-	+	—	+	—
33	+	-	-	+	+	-	+	—	-	+	-	+	-	+
35	+	+	-	-	-	+	+	—	+	-	-	-	+	+
37	-	+	-	+	+	+	-	-	+	-	-	-	+	+
39	-	+	-	+	+	-	+	-	-	-	+	+	+	-
41	+	+	-	-	-	+	-	-	+	+	+	-	-	+
43	+	+	-	-	+	-	+	-	-	-	+	+	-	+
45	+	+	-	+	+	-	-	-	-	-	+	-	+	+
47	—	+	-	-	-	+	+	+	+	-	+	-	+	—
49	+	-	-	+	+	+	-	-	-	-	+	+	-	+
51	+	+	-	-	+	-	+	-	-	+	-	-	+	+
53	+	—	—	+	—	-	+	+	+	-	-	+	+	—
55	+	-	-	-	+	+	+	-	-	+	-	+	-	+
57	+	-	-	+	+	+	-	-	-	+	-	-	+	+
59	+	-	-	+	-	-	-	+	+	+	+	+	-	-
61	+	-	-	-	+	-	+	+	+	-	+	-	-	+
63	-	+	-	+	+	-	+	- 1	+	-	-		+	+

Table 1: Generating odd integers up to 63 by balanced assignments.

Step A ends with a word \underline{x} with $\sigma_0(\underline{x}) = 0$. We turn to Step B and first verify that the shift counter j_B is well-defined.

Lemma 3.1 There is always a cyclic shift of $\langle \underline{x} \rangle_{S \setminus S_0}$ in Step B1 for which $|\sigma_2(\underline{x})| \leq h^2$.

Proof. Let $\underline{x}^{(0)}$ denote the value of \underline{x} at the beginning of Step B1 and let $\underline{x}^{(s)} = (x_{-h}^{(s)}, x_{-h+1}^{(s)}, \dots, x_{h-1}^{(s)})$ be the word obtained from $\underline{x}^{(0)}$ by s right cyclic shifts of $\langle \underline{x}^{(0)} \rangle_{S \setminus S_0}$ (note that $\langle \underline{x}^{(s)} \rangle_{S_0}$ remains zero for all s).

First, we show that $|\sigma_2(\underline{x}^{(s+1)}) - \sigma_2(\underline{x}^{(s)})| \leq 2h^2$ for every $s \geq 0$. We say that location j in $\underline{x}^{(s)}$ contains a sign change if $x_j^{(s)} \neq x_j^{(s+1)}$. Let $j_1 < j_2 < \cdots < j_t$ be the locations of the sign changes in $\underline{x}^{(s)}$. It is easy to verify that

$$\left| \sigma_2(\underline{x}^{(s+1)}) - \sigma_2(\underline{x}^{(s)}) \right| = \left| 2\sum_{i=1}^t (-1)^i \cdot j_i^2 \right|.$$
(3)

Let r be the smallest index i such that $j_i \ge 0$. Define $\beta^- = \sum_{i=1}^{r-1} (-1)^i \cdot j_i^2$ and $\beta^+ = \sum_{i=r}^t (-1)^i \cdot j_i^2$. Now, β^- is a sum of integers with alternating signs and decreasing absolute values, where the first integer in the sum (if any) is negative. Hence,

$$-h^2 \leq -j_1^2 \leq \beta^- \leq 0.$$
 (4)

On the other hand, β^+ is a sum of integers with alternating signs and increasing absolute values. Furthermore, since t is even, the last integer in the sum is positive. Hence,

$$0 \leq \beta^{+} \leq j_{t}^{2} \leq (h-1)^{2} .$$
 (5)

Combining (3), (4), and (5), we obtain,

$$\left| \sigma_2(\underline{x}^{(s+1)}) - \sigma_2(\underline{x}^{(s)}) \right| = 2 \cdot |\beta^- + \beta^+| \le 2h^2 .$$
(6)

Next, we observe that $\sum_{s=0}^{|S\setminus S_0|-1} \sigma_2(\underline{x}^{(s)}) = 0$. Indeed, since $\underline{x}^{(0)}$ is balanced, it follows that $\sum_{s=0}^{|S\setminus S_0|-1} x_j^{(s)} = 0$ for every $j \in S$. Hence,

$$\sum_{s=0}^{|S \setminus S_0|-1} \sigma_2(\underline{x}^{(s)}) = \sum_{s=0}^{|S \setminus S_0|-1} \sum_{j \in S} j^2 \cdot x_j^{(s)} = \sum_{j \in S} j^2 \cdot \sum_{s=0}^{|S \setminus S_0|-1} x_j^{(s)} = 0$$

Therefore, there is a 'zero-crossing' value of s for which $\sigma_2(\underline{x}^{(s)}) \cdot \sigma_2(\underline{x}^{(s+1)}) \leq 0$. By (6), for such an s we must have either $|\sigma_2(\underline{x}^{(s)})| \leq h^2$ or $|\sigma_2(\underline{x}^{(s+1)})| \leq h^2$.

Each iteration in Step B2 changes the value of $\sigma_2(\underline{x})$ by an additive term $\pm (d_i^2 - e_i^2)$, where the negative sign is chosen when $\sigma_2(\underline{x}) \ge 0$. This further reduces the absolute value of $\sigma_2(\underline{x})$ as follows.

Lemma 3.2 The value of $\sigma_2(\underline{x})$ after Step B2 is an odd integer between -63 and 63.

Proof. First note that

$$2(d_{i-1}^2 - e_{i-1}^2) \ge d_i^2 - e_i^2 , \qquad i = 2m - 8, 2m - 9, \dots, 1 , \qquad (7)$$

and $2(d_{2m-8}^2 - e_{2m-8}^2) \ge h^2$. Specifically, for the values in (2) we have $d_i^2 - e_i^2 = 2^{i+6}$ for $i \le 2m-10$, and a simple check reveals that (7) holds also for $i \in 2m-8, 2m-9$ (if $d_{i-1} < -h$, then $\{d_{i-1}, e_{i-1}\}$ is removed from S_{B2} ; nevertheless, it can be verified that (7) still holds if we replace (d_{i-1}, e_{i-1}) by the pair (d_r, e_r) of elements in S_{B2} with the largest index r < i). It follows that after iteration i in Step B2, the resulting absolute value of $\sigma_2(\underline{x})$ is bounded from above by $d_i^2 - e_i^2$. In particular, for i = 0, the value of $\sigma_2(\underline{x})$ is an integer between -64 and 64. Furthermore, at this stage, the only zero entries of \underline{x} are those that are indexed by $S_{B3} \cup S_C$. Since $S \setminus (S_{B3} \cup S_C)$ contains an odd number of odd indexes, it follows that $\sigma_2(\underline{x})$ must be odd.

The final reduction of $|\sigma_2(\underline{x})|$ to zero is done is Step B3, using Table 1. It can be readily checked that for $r = 1, 3, \ldots, 63$, the values in row r in the table contribute r to $\sigma_2(\underline{x})$ (we negate those values in Step B3 if the contribution needs to be -r). Note that neither of the changes made in Step B affects the value of $\sigma_0(\underline{x})$, which still remains zero.

We now turn to Step C. This step is very similar to "Phase A" of the second-order spectral-null encoder in [5, Section IV]). We show next that the swap counter $j_{\rm C}$ is well-defined.

Lemma 3.3 There is always a word \underline{x} obtained by less than h swaps in Step C1 for which $|\sigma_1(\underline{x})| \leq 2(h-1)$.

Proof. Let $\underline{x}^{[0]}$ denote the value of \underline{x} at the beginning of Step C1 and let $\underline{x}^{[j]}$ be the word after the *j*th swap. First, it is easy to check that $|\sigma_1(\underline{x}^{[j+1]}) - \sigma_1(\underline{x}^{[j]})| \leq 4(h-1)$ for all $j \geq 0$. Suppose we continue the swaps until j = h-1, and let $\underline{x}^{[h]}$ be the word obtained from $\underline{x}^{[h-1]}$ by negating the first entry (indexed by -h). In that case we will have $|\sigma_1(\underline{x}^{[h]}) - \sigma_1(\underline{x}^{[h-1]})| = 2h$ and

$$\sigma_1(\underline{x}^{[h]}) = -\sigma_1(\underline{x}^{[0]}) .$$

Hence, there must be a 'zero-crossing' index j < h for which $\sigma_1(\underline{x}^{[j]}) \cdot \sigma_1(\underline{x}^{[j+1]}) \leq 0$. For such a j we must have either $|\sigma_1(\underline{x}^{[j]})| \leq 2(h-1)$ or $|\sigma_1(\underline{x}^{[j+1]})| \leq 2(h-1)$. Furthermore, if the zero-crossing index is j = h-1, we have $|\sigma_1(\underline{x}^{[h-1]})| \leq h$ or $|\sigma_1(\underline{x}^{[h]})| = |\sigma_1(\underline{x}^{[0]})| \leq h$.

Turning to Step C2, it can be easily verified that after iteration i in that step, the resulting value of $|\sigma_1(\underline{x})|$ is bounded from above by 2^{i+1} . In particular, for i = 0, the value of $\sigma_1(\underline{x})$ is an integer between -2 and 2. The following lemma implies that $\sigma_1(\underline{x})$ is actually zero at this point.

Lemma 3.4 For *n* divisible by 4 and every $\underline{w} \in F^n$,

$$\sigma_1(\underline{w}) \equiv \sigma_2(\underline{w}) \pmod{4}$$
.

Proof. Let n = 2h and write $\underline{w} = (w_{-h}, w_{-h+1}, \ldots, w_{h-1})$. Then,

$$\begin{aligned} \sigma_2(\underline{w}) - \sigma_1(\underline{w}) &= \sum_{j=-h}^{h-1} j(j-1) \cdot w_j \\ &= \sum_{l=-h/2}^{(h/2)-1} \left((2l)(2l-1) \cdot w_{2l} + (2l+1)(2l) \cdot w_{2l+1} \right) \\ &= \sum_{l=-h/2}^{(h/2)-1} (2l) \left((2l-1) \cdot w_{2l} + (2l+1) \cdot w_{2l+1} \right) . \end{aligned}$$

The result follows by observing that $(2l)((2l-1) \cdot w_{2l} + (2l+1) \cdot w_{2l+1})$ is divisible by 4 for every l.

Neither of the changes made in Step C affects the values of $\sigma_0(\underline{x})$ or $\sigma_2(\underline{x})$, which still remain zero. Hence, at the end of Step C we will have $\sigma_1(\underline{x}) \equiv 0 \pmod{4}$. And since $-2 \leq \sigma_1(\underline{x}) \leq 2$, it follows that $\sigma_1(\underline{x})$ is zero.

Finally, Step D is rather straightforward and is based on the fact that the concatenation of two k-th order spectral-null words yields a k-th order spectral-null word.

Decoding of \underline{y} is done by first reconstructing the values $j_{\rm B}$ and $j_{\rm C}$ from \underline{x}' . Once we have those two counters, we can reconstruct the values of \underline{x} at the beginning of Steps C and B (in that order).

3.2 Redundancy

We now compute the redundancy of the code which is defined by the words generated by the algorithm for any given length.

Using the algorithms in [1], [3], or [8], the redundancy in Step A due to the balancing of y is at most m bits.

Steps B and C require $|S_0| \leq 6m-2$ bits to reduce $|\sigma_2(\underline{x})|$ and $|\sigma_1(\underline{x})|$ to zero. We also need m bits to represent the shift counter $j_{\rm B}$ and m-1 bits to represent the swap counter $j_{\rm C}$.

In Step D, the encoding procedure is applied recursively to the 2m-1 bits that represent $(j_{\rm B}, j_{\rm C})$, thus generating a word $\underline{x}' \in \mathcal{S}(m', 3)$ of length $m' = 2m + O(\log m)$. Since $m = \lceil \log_2 n \rceil$, it follows that the total redundancy of the encoding scheme is $9 \log_2 n + O(\log \log n)$ bits. This expression will be an upper bound on the redundancy also if we replace n by the overall length, n + m', of the output word.

3.3 Time and space complexity

Step A can be implemented by O(n) increments/decrements of a $\lceil \log_2 n \rceil$ -bit counter, and a look-up table of size $\lceil \log_2 n \rceil^2$ bits.

As for Step B, we need to have the value of $\sigma_2(\underline{x})$ for each cyclic shift in Step B1. Assuming that the squares of the elements between 1 and h are pre-computed in a table, the initial value of $\sigma_2(\underline{x})$ in this step can be found in O(n) additions of $O(\log n)$ -bit integers. Now, let $\underline{\hat{x}}$ denote the word obtained from \underline{x} by one right cyclic shift of $\langle \underline{x} \rangle_{S \setminus S_0}$, and let $\underline{\tilde{x}}$ be the word obtained from \underline{x} by one right cyclic shift of the whole word \underline{x} . We describe next how $\sigma_{\ell}(\underline{\hat{x}})$ can be computed efficiently from $\sigma_{\ell}(\underline{x})$, $\ell = 1, 2$. Step B1 will then proceed iteratively by making $\underline{\hat{x}}$ the new value of \underline{x} . Noting that $\sigma_0(\underline{x}) = 0$, it is easy to verify that

$$\sigma_1(\underline{\tilde{x}}) = \sigma_1(\underline{x}) - 2h \cdot x_{h-1}$$
 and $\sigma_2(\underline{\tilde{x}}) = \sigma_2(\underline{x}) + 2\sigma_1(\underline{x})$

Therefore, once we have $\sigma_1(\underline{x})$ and $\sigma_2(\underline{x})$, it is straightforward to compute $\sigma_1(\underline{\tilde{x}})$ and $\sigma_2(\underline{\tilde{x}})$.

Let S_1 denote the set of all indexes $j \in S_0$ such that $j-1 \in S \setminus S_0$ (when j = -h, the index j-1 should read h-1). For an index $j \in S_1$, let \hat{j} denote the smallest index in $S \setminus S_0$ that is larger than j (if no such index exists, then \hat{j} is defined as the smallest index in $S \setminus S_0$). For $\ell = 1, 2$, define

$$\alpha_\ell(\underline{x}) = \sum_{j \in S_1} (\hat{j}^\ell - j^\ell) \cdot x_{j-1} \; .$$

It can be readily verified that

$$\sigma_{\ell}(\underline{\hat{x}}) = \sigma_{\ell}(\underline{\tilde{x}}) + \alpha_{\ell}(\underline{x}) , \qquad \ell = 1, 2 .$$

The expressions $\alpha_{\ell}(\underline{x})$ can be computed using $O(\log n)$ additions of $O(\log n)$ -bit integers. The following discussion outlines how the computation of $\alpha_{\ell}(\underline{x})$ can be accelerated further through the use of small look-up tables.

Let $S_1 = \bigcup_t S_1(t)$ be a partition of S_1 into O(1) subsets $S_1(t)$, each of size less than m. For each subset $S_1(t)$, construct a look-up table for computing the expression

$$\alpha_{\ell}\left(\langle \underline{x} \rangle_{S_1(t)}\right) = \sum_{j \in S_1(t)} (\hat{j}^{\ell} - j^{\ell}) \cdot x_{j-1} ,$$

as a function of the entries $x_{j-1}, j \in S_1(t)$. Each look-up table consists of less than n entries and each entry contains an $O(\log n)$ -bit integer. Note that these look-up tables can be computed in time O(n) and that they depend on n, but not on the encoded word. In order to access the bits $x_{j-1}, j \in S_1(t)$ within $\langle \underline{x} \rangle_{S \setminus S_0}$, we will use $|S_1(t)|$ pointers (counters) that will be decremented after each shift. (In hardware implementations, we can instead store $\langle \underline{x} \rangle_{S \setminus S_0}$ in a shift-register.) Once we have computed the O(1) expressions $\alpha_{\ell}(\langle \underline{x} \rangle_{S_1(t)})$, we obtain $\alpha_{\ell}(\underline{x})$ as their sum. Note that this computation of $\sigma_2(\underline{x})$ allows us to find $j_{\rm B}$ without actually shifting $\langle \underline{x} \rangle_{S \setminus S_0}$. This makes the computation efficient also in software implementations.

Steps B2, B3, and C are rather straightforward and can be implemented using O(n) integer additions. Hence, the overall time and space complexity of our encoding algorithm is as follows:

- O(n) additions of $O(\log n)$ -bit integers,
- O(n) accesses to O(1) tables, each of size < n, and —
- O(n) increments/decrements of $O(\log n)$ counters, each $\lceil \log_2 n \rceil$ bits long.

4 Example

We consider here the case n = 60 (for such a small value of n the redundancy is relatively big, so this example is given only for the purpose of illustrating the encoding steps). In this case h = 30 and m = 6, and the set S_{B2} is given by

$$S_{\rm B2} = \{-10, -18, -20, -23\} \cup \{-6, -14, -12, 7\},\$$

where $\tau_1 = 23$. Note that we have excluded the elements $\{d_3, e_3\} = \{\tau_1, \tau_2\} = \{23, 15\}$ from S_{B2} since they will not be required in Step B2: The value of $d_2^2 - e_2^2 = (-20)^2 - (-12)^2 = 256$ is already greater than half the value of $d_4^2 - e_4^2 = (-23)^2 - 7^2 = 480$. The set S_{B3} is of size 14 and S_C is given by $\{\pm 2^i\}_{i=0}^4$. Hence, $|S_0| = 32$.

Suppose that the input balanced word y of length $n - |S_0| = 28$ is given by

After embedding y in \underline{x} in Step A, we obtain the word

(the arrow points at the entry indexed by 0). For this word we have $\sigma_0(\underline{x}) = 0$ and $\sigma_2(\underline{x}) = -2047$, and when applying the cyclic shifts in Step B1 we produce words \underline{x} with $\sigma_2(\underline{x}) = -1853, -1755, -1357$, and -625. The last value corresponds to

which is the first word in this step with $|\sigma_2(\underline{x})| \leq h^2 = 900$; so, $j_B = 4$. The assignment of values to the entries indexed by S_{B2} in Step B2 results in

with $\sigma_2(\underline{x}) = 47$. In Step B3, we fill in the entries indexed by S_{B3} with the negated entries of the row that corresponds to 47 in Table 1. This produces the word

Step C1 starts with $\sigma_1(\underline{x}) = 174$ and then continues with iterated swaps that generate words \underline{x} with $\sigma_1(\underline{x}) = 194, 194, 182$ (9 iterations), 134, 82, 82, and 22. The last value corresponds to

and this word is the first to occur in this step with $|\sigma_1(\underline{x})| \leq 2(h-1) = 58$, and so $j_{\rm C} = 15$. Step C2 fills in the entries indexed by $S_{\rm C}$ to produce the word

for which we have $\sigma_0(\underline{x}) = \sigma_1(\underline{x}) = \sigma_2(\underline{x}) = 0$.

Note that we can make the counting of the swaps in Step C more economical by skipping index pairs (-j, j) with $x_{-j} = x_j$ (in which case the swaps become in effect negations of x_{-j} and x_j whenever $x_{-j} \neq x_j$).

Finally, the counters $(j_{\rm B}, j_{\rm C})$ are coded into up to $2 \cdot 6 - 1 = 11$ bits and undergo a recursive encoding in Step D.

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